International TOR Rectifier

IRF7807VD2PbF

FETKY™ MOSFET / SCHOTTKY DIODE

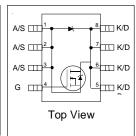
- Co-Pack N-channel HEXFET[®] Power MOSFET and Schottky Diode
- Ideal for Synchronous Rectifiers in DC-DC Converters Up to 5A Output
- Low Conduction Losses
- Low Switching Losses
- · Low Vf Schottky Rectifier
- · Lead-Free

Description

The FETKY™ family of Co-Pack HEXFET®MOSFETs and Schottky diodes offers the designer an innovative, board space saving solution for switching regulator and power management applications. HEXFET power MOSFETs utilize advanced processing techniques to achieve extremely low on-resistance per silicon area. Combining this technology with International Rectifier's low forward drop Schottky rectifiers results in an extremely efficient device suitable for use in a wide variety of portable electronics applications.

The SO-8 has been modified through a customized leadframe for enhanced thermal characteristics. The SO-8 package is designed for vapor phase, infrared or wave soldering techniques.





DEVICE CHARACTERISTICS®

	IRF7807VD2					
$R_{DS(on)}$	17mΩ					
Q_{G}	9.5nC					
Q _{sw}	3.4nC					
Q _{oss}	12nC					

Absolute Maximum Ratings

About Maximum Radingo							
Parameter		Symbol	Max.	Units			
Drain-Source Voltage		V _{DS}	30	V			
Gate-Source Voltage		V _{GS}	±20	V			
Continuous Drain or Source	25°C	I _D	8.3				
Current ($V_{GS} \ge 4.5V$)	70°C		6.6	А			
Pulsed Drain Current®		I _{DM}	66				
Power Dissipation® 25°C		P _D	2.5	W			
	70°C		1.6	VV			
Schottky and Body Diode	25°C	I _F (AV)	3.7	А			
Average ForwardCurrent@	70°C		2.3				
Junction & Storage Temperatu	ire Range	T_J, T_{STG}	-55 to 150	°C			

Thermal Resistance

Parameter		Max.	Units
Maximum Junction-to-Ambient®	$R_{_{\thetaJA}}$	50	°C/W
Maximum Junction-to-Lead	$R_{_{\theta JL}}$	20	°C/W

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Electrical Characteristics

Parameter		Min	Тур	Max	Units	Conditions
Drain-to-Source Breakdown Voltage	BV _{DSS}	30	-	-	V	$V_{GS} = 0V, I_{D} = 250\mu A$
Static Drain-Source on Resistance	R _{DS(on)}		17	25	m $Ω$	$V_{GS} = 4.5V, I_{D} = 7.0A@$
Gate Threshold Voltage	$V_{GS(th)}$	1.0			V	$V_{DS} = V_{GS}, I_{D} = 250\mu A$
Drain-Source Leakage	I _{DSS}			50	μΑ	$V_{DS} = 24V, V_{GS} = 0$
Current				6.0	mA	$V_{DS} = 24V, V_{GS} = 0,$
						Tj = 100°C
Gate-Source Leakage Current*	I _{GSS}			±100	nA	$V_{GS} = \pm 20V$
Total Gate Charge*	Q_{G}		9.5	14		V_{GS} =4.5V, I_{D} =7.0A
Pre-Vth Gate-Source Charge	Q _{GS1}		2.3			V _{DS} = 16V
Post-Vth Gate-Source Charge	Q _{GS2}		1.0		nC	
Gate to Drain Charge	Q_{GD}		2.4			
Switch Chg(Q _{qs2} + Q _{qd})	Q_{sw}		3.4	5.2		
Output Charge*	Q _{oss}		12	16.8		$V_{DS} = 16V, V_{GS} = 0$
Gate Resistance	R_{G}		2.0		Ω	
Turn-on Delay Time	t _{d (on)}		6.3			$V_{DD} = 16V, I_{D} = 7.0A$
Rise Time	t _r		1.2		ns	$V_{GS} = 5V$, $R_{G} = 2\Omega$
Turn-off Delay Time	t _{d (off)}		11			Resistive Load
Fall Time	t,		2.2			

Schottky Diode & Body Diode Ratings and Characteristics

Parameter		Min	Тур	Max	Units	Conditions
Diode Forward Voltage	V _{SD}			0.54	V	$T_i = 25$ °C, $I_s = 3.0$ A, $V_{GS} = 0$ V2
				0.43	[$T_j = 125$ °C, $I_s = 3.0$ A, $V_{GS} = 0$ V
Reverse Recovery Time	trr		36		ns	$T_{j} = 25^{\circ}C$, $I_{s} = 7.0A$, $V_{DS} = 16V$
Reverse Recovery Charge	Qrr		41		nC	di/dt = 100A/µs
Forward Turn-On Time	t _{on}	Intrinsic turn-on time is negligible (turn-on is dominated by L _s +L _D)				

- Notes:
 Repetitive rating; pulse width limited by max. junction temperature. Pulse width $\leq 400 \ \mu s$; duty cycle $\leq 2\%$.

 When mounted on 1 inch square copper board

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Power MOSFET Selection for DC/DC Converters

Control FET

Special attention has been given to the power losses in the switching elements of the circuit - Q1 and Q2. Power losses in the high side switch Q1, also called the Control FET, are impacted by the $R_{\rm ds(on)}$ of the MOSFET, but these conduction losses are only about one half of the total losses.

Power losses in the control switch Q1 are given by;

$$P_{loss} = P_{conduction} + P_{switching} + P_{drive} + P_{output}$$

This can be expanded and approximated by;

$$\begin{split} P_{loss} &= \left(I_{rms}^{2} \times R_{ds(on)}\right) \\ &+ \left(I \times \frac{Q_{gd}}{i_{g}} \times V_{in} \times f\right) + \left(I \times \frac{Q_{gs2}}{i_{g}} \times V_{in} \times f\right) \\ &+ \left(Q_{g} \times V_{g} \times f\right) \\ &+ \left(\frac{Q_{oss}}{2} \times V_{in} \times f\right) \end{split}$$

This simplified loss equation includes the terms Q_{gs2} and Q_{oss} which are new to Power MOSFET data sheets.

 Q_{gs2} is a sub element of traditional gate-source charge that is included in all MOSFET data sheets. The importance of splitting this gate-source charge into two sub elements, Q_{gs1} and Q_{gs2} , can be seen from Fig. 1

 Q_{gs2} indicates the charge that must be supplied by the gate driver between the time that the threshold voltage has been reached (t1) and the time the drain current rises to I_{dmax} (t2) at which time the drain voltage begins to change. Minimizing Q_{gs2} is a critical factor in reducing switching losses in Q1.

 $Q_{_{oss}}$ is the charge that must be supplied to the output capacitance of the MOSFET during every switching cycle. Figure 2 shows how $Q_{_{oss}}$ is formed by the parallel combination of the voltage dependant (nonlinear) capacitance's $C_{_{ds}}$ and $C_{_{dg}}$ when multiplied by the power supply input buss voltage.

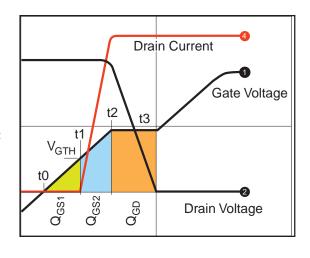


Figure 1: Typical MOSFET switching waveform

Synchronous FET

The power loss equation for Q2 is approximated by:

$$\begin{aligned} P_{loss} &= P_{conduction} + P_{drive} + P_{output}^* \\ P_{loss} &= \left(I_{rms}^2 \times R_{ds(on)}\right) \\ &+ \left(Q_g \times V_g \times f\right) \\ &+ \left(\frac{Q_{oss}}{2} \times V_{in} \times f\right) + \left(Q_{rr} \times V_{in} \times f\right) \end{aligned}$$

*dissipated primarily in Q1.

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For the synchronous MOSFET Q2, $R_{\rm ds(on)}$ is an important characteristic; however, once again the importance of gate charge must not be overlooked since it impacts three critical areas. Under light load the MOSFET must still be turned on and off by the control IC so the gate drive losses become much more significant. Secondly, the output charge $Q_{\rm oss}$ and reverse recovery charge $Q_{\rm r}$ both generate losses that are transfered to Q1 and increase the dissipation in that device. Thirdly, gate charge will impact the MOSFETs' susceptibility to Cdv/dt turn on.

The drain of Q2 is connected to the switching node of the converter and therefore sees transitions between ground and $V_{\rm in}$. As Q1 turns on and off there is a rate of change of drain voltage dV/dt which is capacitively coupled to the gate of Q2 and can induce a voltage spike on the gate that is sufficient to turn

the MOSFET on, resulting in shoot-through current . The ratio of $Q_{\rm gd}/Q_{\rm gs1}$ must be minimized to reduce the potential for Cdv/dt turn on.

Spice model for IRF7807V can be downloaded in machine readable format at www.irf.com.

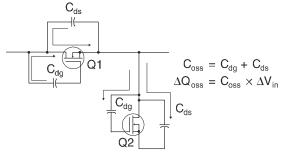
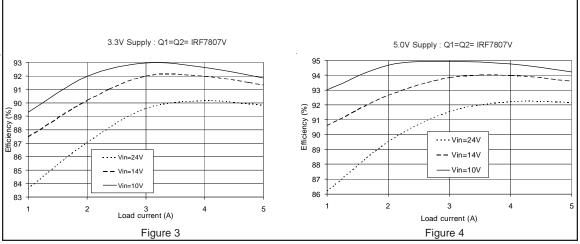


Figure 2: Qoss Characteristic

Typical Mobile PC Application

The performance of these new devices has been tested in circuit and correlates well with performance predictions generated by the system models. An advantage of this new technology platform is that the MOSFETs it produces are suitable for both control FET and synchronous FET applications. This has been demonstrated with the 3.3V and 5V converters. (Fig 3 and Fig 4). In these applications the same MOSFET IRF7807V was used for both the control FET (Q1) and the synchronous FET (Q2). This provides a highly effective cost/performance solution.



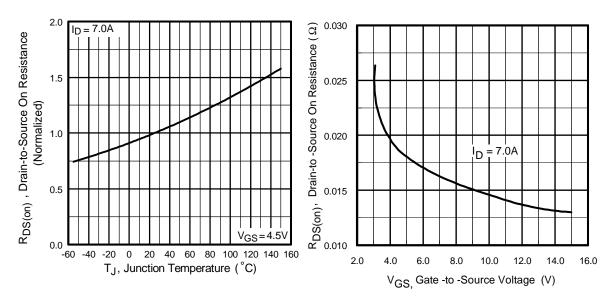


Fig 5. Normalized On-Resistance Vs. Temperature

Fig 7. On-Resistance Vs. Gate Voltage

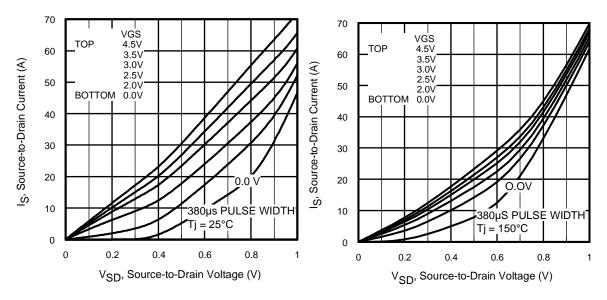


Fig 7. Typical Reverse Output Characteristics

Fig 8. Typical Reverse Output Characteristics

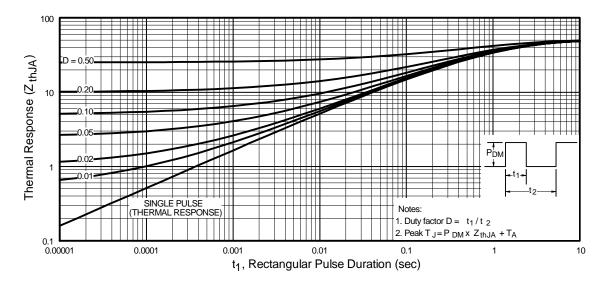


Figure 9. Maximum Effective Transient Thermal Impedance, Junction-to-Ambient

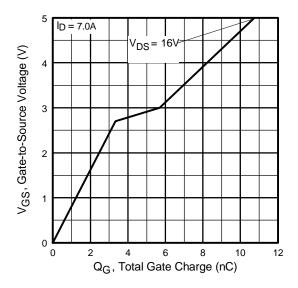
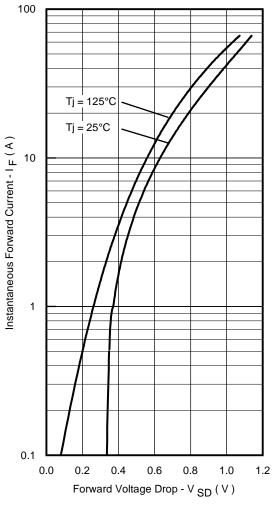


Fig 10. Typical Gate Charge Vs. Gate-to-Source Voltage

MOSFET, Body Diode & Schottky Diode Characteristics



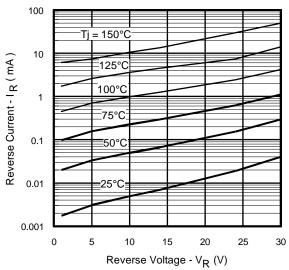


Fig. 12 - Typical Values of Reverse Current Vs. Reverse Voltage

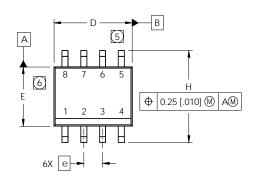
Fig. 11 - Typical Forward Voltage Drop Characteristics

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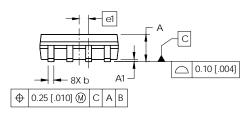
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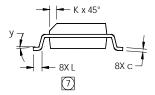
SO-8 (Fetky) Package Outline

Dimensions are shown in millimeters (inches)



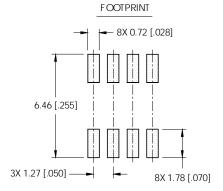
DIM	INC	HES	MILLIMETERS		
DIIVI	MIN	MAX	MIN	MAX	
Α	.0532	.0688	1.35	1.75	
A1	.0040	.0098 0.10		0.25	
b	.013	.020	0.33	0.51	
С	.0075	.0098	0.19	0.25	
D	.189	.1968	4.80	5.00	
Ε	.1497	.1574	3.80	4.00	
е	.050 B	ASIC	1.27 BASIC		
e1	.025 B	ASIC	0.635 BASIC		
Н	.2284	.2440	5.80	6.20	
K	.0099	.0196	0.25	0.50	
L	.016	.050	0.40	1.27	
у	0°	8°	0°	8°	



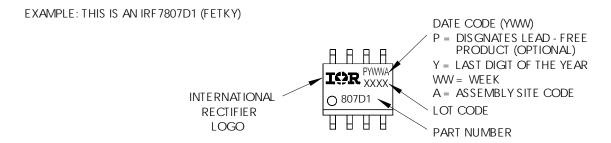


NOTES:

- 1. DIMENSIONING & TOLERANCING PER ASME Y14.5M-1994.
- 2. CONTROLLING DIMENSION: MILLIMETER
- 3. DIMENSIONS ARE SHOWN IN MILLIMETERS [INCHES].
- 4. OUTLINE CONFORMS TO JEDEC OUTLINE MS-012AA.
- (5) DIMENSION DOES NOT INCLUDE MOLD PROTRUSIONS. MOLD PROTRUSIONS NOT TO EXCEED 0.15 [.006].
- (6) DIMENSION DOES NOT INCLUDE MOLD PROTRUSIONS. MOLD PROTRUSIONS NOT TO EXCEED 0.25 [.010].
- (7) DIMENSION IS THE LENGTH OF LEAD FOR SOLDERING TO A SUBSTRATE.

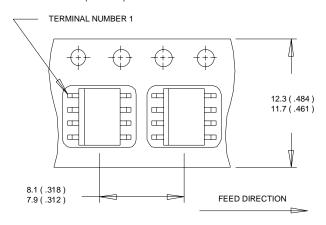


SO-8 (Fetky) Part Marking Information



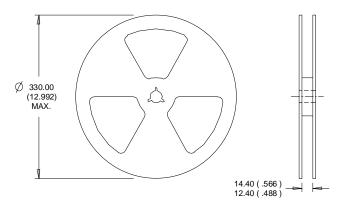
SO-8 (Fetky) Tape and Reel

Dimensions are shown in millimeters (inches)



NOTES:

- CONTROLLING DIMENSION : MILLIMETER.
 ALL DIMENSIONS ARE SHOWN IN MILLIMETERS(INCHES).
- 3. OUTLINE CONFORMS TO EIA-481 & EIA-541.



- NOTES:
 1. CONTROLLING DIMENSION: MILLIMETER.
 2. OUTLINE CONFORMS TO EIA-481 & EIA-541.

Data and specifications subject to change without notice. This product has been designed and qualified for the Consumer market. Qualifications Standards can be found on IR's Web site.

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